

Fundamentals of Applied Electromagnetics 6e  
by  
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Tables

# Chapters

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## Chapter 1 Tables

**Table 1-1** Fundamental SI units.

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**Table 1-1:** Fundamental SI units.

<b>Dimension</b>	<b>Unit</b>	<b>Symbol</b>
<b>Length</b>	meter	m
<b>Mass</b>	kilogram	kg
<b>Time</b>	second	s
<b>Electric Current</b>	ampere	A
<b>Temperature</b>	kelvin	K
<b>Amount of substance</b>	mole	mol

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**Table 1-2:** Multiple and submultiple prefixes.

Prefix	Symbol	Magnitude
<b>exa</b>	E	$10^{18}$
<b>peta</b>	P	$10^{15}$
<b>tera</b>	T	$10^{12}$
<b>giga</b>	G	$10^9$
<b>mega</b>	M	$10^6$
<b>kilo</b>	k	$10^3$
<b>milli</b>	m	$10^{-3}$
<b>micro</b>	$\mu$	$10^{-6}$
<b>nano</b>	n	$10^{-9}$
<b>pico</b>	p	$10^{-12}$
<b>femto</b>	f	$10^{-15}$
<b>atto</b>	a	$10^{-18}$

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**Table 1-3:** The three branches of electromagnetics.

<b>Branch</b>	<b>Condition</b>	<b>Field Quantities (Units)</b>
<b>Electrostatics</b>	Stationary charges ( $\partial q/\partial t = 0$ )	Electric field intensity <b>E</b> (V/m) Electric flux density <b>D</b> (C/m <sup>2</sup> ) <b>D</b> = $\epsilon\mathbf{E}$
<b>Magnetostatics</b>	Steady currents ( $\partial I/\partial t = 0$ )	Magnetic flux density <b>B</b> (T) Magnetic field intensity <b>H</b> (A/m) <b>B</b> = $\mu\mathbf{H}$
<b>Dynamics</b> <b>(Time-varying fields)</b>	Time-varying currents ( $\partial I/\partial t \neq 0$ )	<b>E, D, B, and H</b> ( <b>E, D</b> ) coupled to ( <b>B, H</b> )

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**Table 1-4:** Constitutive parameters of materials.

Parameter	Units	Free-space Value
<b>Electrical permittivity <math>\epsilon</math></b>	F/m	$\epsilon_0 = 8.854 \times 10^{-12}$ (F/m) $\simeq \frac{1}{36\pi} \times 10^{-9}$ (F/m)
<b>Magnetic permeability <math>\mu</math></b>	H/m	$\mu_0 = 4\pi \times 10^{-7}$ (H/m)
<b>Conductivity <math>\sigma</math></b>	S/m	0

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**Table 1-5:** Time-domain sinusoidal functions  $z(t)$  and their cosine-reference phasor-domain counterparts  $\tilde{Z}$ , where  $z(t) = \Re\{\tilde{Z}e^{j\omega t}\}$ .

$z(t)$		$\tilde{Z}$
$A \cos \omega t$	$\longleftrightarrow$	$A$
$A \cos(\omega t + \phi_0)$	$\longleftrightarrow$	$Ae^{j\phi_0}$
$A \cos(\omega t + \beta x + \phi_0)$	$\longleftrightarrow$	$Ae^{j(\beta x + \phi_0)}$
$Ae^{-\alpha x} \cos(\omega t + \beta x + \phi_0)$	$\longleftrightarrow$	$Ae^{-\alpha x} e^{j(\beta x + \phi_0)}$
$A \sin \omega t$	$\longleftrightarrow$	$Ae^{-j\pi/2}$
$A \sin(\omega t + \phi_0)$	$\longleftrightarrow$	$Ae^{j(\phi_0 - \pi/2)}$
$\frac{d}{dt}(z(t))$	$\longleftrightarrow$	$j\omega\tilde{Z}$
$\frac{d}{dt}[A \cos(\omega t + \phi_0)]$	$\longleftrightarrow$	$j\omega A e^{j\phi_0}$
$\int z(t) dt$	$\longleftrightarrow$	$\frac{1}{j\omega}\tilde{Z}$
$\int A \sin(\omega t + \phi_0) dt$	$\longleftrightarrow$	$\frac{1}{j\omega} A e^{j(\phi_0 - \pi/2)}$



## Chapter 2 Tables

**Table 2-1** Transmission-line parameters  $R'$ ,  $L'$ ,  $G'$ , and  $C'$  for three types of lines.

**Table 2-2** Characteristic parameters of transmission lines.

**Table 2-3** Magnitude and phase of the reflection coefficient for various types of loads.

**Table 2-4** Properties of standing waves on a lossless transmission line.

**Table 2-1:** Transmission-line parameters  $R'$ ,  $L'$ ,  $G'$ , and  $C'$  for three types of lines.

Parameter	Coaxial	Two-Wire	Parallel-Plate	Unit
$R'$	$\frac{R_s}{2\pi} \left( \frac{1}{a} + \frac{1}{b} \right)$	$\frac{2R_s}{\pi d}$	$\frac{2R_s}{w}$	$\Omega/\text{m}$
$L'$	$\frac{\mu}{2\pi} \ln(b/a)$	$\frac{\mu}{\pi} \ln \left[ (D/d) + \sqrt{(D/d)^2 - 1} \right]$	$\frac{\mu h}{w}$	$\text{H}/\text{m}$
$G'$	$\frac{2\pi\sigma}{\ln(b/a)}$	$\frac{\pi\sigma}{\ln \left[ (D/d) + \sqrt{(D/d)^2 - 1} \right]}$	$\frac{\sigma w}{h}$	$\text{S}/\text{m}$
$C'$	$\frac{2\pi\epsilon}{\ln(b/a)}$	$\frac{\pi\epsilon}{\ln \left[ (D/d) + \sqrt{(D/d)^2 - 1} \right]}$	$\frac{\epsilon w}{h}$	$\text{F}/\text{m}$







Notes: (1) Refer to Fig. ?? for definitions of dimensions. (2)  $\mu$ ,  $\epsilon$ , and  $\sigma$  pertain to the insulating material between the conductors. (3)  $R_s = \sqrt{\pi f \mu_c / \sigma_c}$ . (4)  $\mu_c$  and  $\sigma_c$  pertain to the conductors. (5) If  $(D/d)^2 \gg 1$ , then  $\ln \left[ (D/d) + \sqrt{(D/d)^2 - 1} \right] \simeq \ln(2D/d)$ .

**Table 2-2:** Characteristic parameters of transmission lines.

	<b>Propagation Constant</b> $\gamma = \alpha + j\beta$	<b>Phase Velocity</b> $u_p$	<b>Characteristic Impedance</b> $Z_0$
<b>General case</b>	$\gamma = \sqrt{(R' + j\omega L')(G' + j\omega C')}$	$u_p = \omega/\beta$	$Z_0 = \sqrt{\frac{(R' + j\omega L')}{(G' + j\omega C')}}$
<b>Lossless</b> ( $R' = G' = 0$ )	$\alpha = 0, \beta = \omega\sqrt{\epsilon_r}/c$	$u_p = c/\sqrt{\epsilon_r}$	$Z_0 = \sqrt{L'/C'}$
<b>Lossless coaxial</b>	$\alpha = 0, \beta = \omega\sqrt{\epsilon_r}/c$	$u_p = c/\sqrt{\epsilon_r}$	$Z_0 = (60/\sqrt{\epsilon_r}) \ln(b/a)$
<b>Lossless two-wire</b>	$\alpha = 0, \beta = \omega\sqrt{\epsilon_r}/c$	$u_p = c/\sqrt{\epsilon_r}$	$Z_0 = (120/\sqrt{\epsilon_r}) \cdot \ln[(D/d) + \sqrt{(D/d)^2 - 1}]$ $Z_0 \simeq (120/\sqrt{\epsilon_r}) \ln(2D/d),$ if $D \gg d$
<b>Lossless parallel-plate</b>	$\alpha = 0, \beta = \omega\sqrt{\epsilon_r}/c$	$u_p = c/\sqrt{\epsilon_r}$	$Z_0 = (120\pi/\sqrt{\epsilon_r})(h/w)$

Notes: (1)  $\mu = \mu_0$ ,  $\epsilon = \epsilon_r \epsilon_0$ ,  $c = 1/\sqrt{\mu_0 \epsilon_0}$ , and  $\sqrt{\mu_0/\epsilon_0} \simeq (120\pi) \Omega$ , where  $\epsilon_r$  is the relative permittivity of insulating material. (2) For coaxial line,  $a$  and  $b$  are radii of inner and outer conductors. (3) For two-wire line,  $d$  = wire diameter and  $D$  = separation between wire centers. (4) For parallel-plate line,  $w$  = width of plate and  $h$  = separation between the plates.

**Table 2-3:** Magnitude and phase of the reflection coefficient for various types of loads. In general,  $z_L = Z_L/Z_0 = (R + jX)/Z_0 = r + jx$ , where  $r = R/Z_0$  and  $x = X/Z_0$  are the real and imaginary parts of the normalized load impedance  $z_L$ , respectively.

Load	$Z_L$	Reflection Coefficient $\Gamma =  \Gamma e^{j\theta_r}$	$ \Gamma $	$\theta_r$
	$Z_L = (r + jx)Z_0$		$\left[ \frac{(r-1)^2 + x^2}{(r+1)^2 + x^2} \right]^{1/2}$	$\tan^{-1} \left( \frac{x}{r-1} \right) - \tan^{-1} \left( \frac{x}{r+1} \right)$
	$Z_0$	0 (no reflection)	0	irrelevant
	(short)	1	1	$\pm 180^\circ$ (phase opposition)
	(open)	1	1	0 (in-phase)
	$jX = j\omega L$	1	1	$\pm 180^\circ - 2 \tan^{-1} x$
	$jX = \frac{-j}{\omega C}$	1	1	$\pm 180^\circ + 2 \tan^{-1} x$

**Table 2-4:** Properties of standing waves on a lossless transmission line.

<b>Voltage maximum</b>	$ \tilde{V} _{\max} =  V_0^+  [1 +  \Gamma ]$
<b>Voltage minimum</b>	$ \tilde{V} _{\min} =  V_0^+  [1 -  \Gamma ]$
Positions of voltage maxima (also positions of current minima)	$d_{\max} = \frac{\theta_r \lambda}{4\pi} + \frac{n\lambda}{2}, \quad n = 0, 1, 2, \dots$
Position of first maximum (also position of first current minimum)	$d_{\max} = \begin{cases} \frac{\theta_r \lambda}{4\pi}, & \text{if } 0 \leq \theta_r \leq \pi \\ \frac{\theta_r \lambda}{4\pi} + \frac{\lambda}{2}, & \text{if } -\pi \leq \theta_r \leq 0 \end{cases}$
Positions of voltage minima (also positions of current maxima)	$d_{\min} = \frac{\theta_r \lambda}{4\pi} + \frac{(2n+1)\lambda}{4}, \quad n = 0, 1, 2, \dots$
Position of first minimum (also position of first current maximum)	$d_{\min} = \frac{\lambda}{4} \left( 1 + \frac{\theta_r}{\pi} \right)$
<b>Input impedance</b>	$Z_{\text{in}} = Z_0 \left( \frac{z_L + j \tan \beta l}{1 + j z_L \tan \beta l} \right) = Z_0 \left( \frac{1 + \Gamma_l}{1 - \Gamma_l} \right)$
Positions at which $Z_{\text{in}}$ is real	at voltage maxima and minima
$Z_{\text{in}}$ at voltage maxima	$Z_{\text{in}} = Z_0 \left( \frac{1 +  \Gamma }{1 -  \Gamma } \right)$
$Z_{\text{in}}$ at voltage minima	$Z_{\text{in}} = Z_0 \left( \frac{1 -  \Gamma }{1 +  \Gamma } \right)$
$Z_{\text{in}}$ of short-circuited line	$Z_{\text{in}}^{\text{sc}} = jZ_0 \tan \beta l$
$Z_{\text{in}}$ of open-circuited line	$Z_{\text{in}}^{\text{oc}} = -jZ_0 \cot \beta l$
$Z_{\text{in}}$ of line of length $l = n\lambda/2$	$Z_{\text{in}} = Z_L, \quad n = 0, 1, 2, \dots$
$Z_{\text{in}}$ of line of length $l = \lambda/4 + n\lambda/2$	$Z_{\text{in}} = Z_0^2/Z_L, \quad n = 0, 1, 2, \dots$
$Z_{\text{in}}$ of matched line	$Z_{\text{in}} = Z_0$
$ V_0^+ $ = amplitude of incident wave; $\Gamma =  \Gamma e^{j\theta_r}$ with $-\pi < \theta_r < \pi$ ; $\theta_r$ in radians; $\Gamma_l = \Gamma e^{-j2\beta l}$ .	

## Chapter 3 Tables

**Table 3-1** Summary of vector relations.

**Table 3-2** Coordinate transformation relations.

**Table 3-1:** Summary of vector relations.

	Cartesian Coordinates	Cylindrical Coordinates	Spherical Coordinates
<b>Coordinate variables</b>	$x, y, z$	$r, \phi, z$	$R, \theta, \phi$
<b>Vector representation</b> $\mathbf{A} =$	$\hat{\mathbf{x}}A_x + \hat{\mathbf{y}}A_y + \hat{\mathbf{z}}A_z$	$\hat{\mathbf{r}}A_r + \hat{\boldsymbol{\phi}}A_\phi + \hat{\mathbf{z}}A_z$	$\hat{\mathbf{R}}A_R + \hat{\boldsymbol{\theta}}A_\theta + \hat{\boldsymbol{\phi}}A_\phi$
<b>Magnitude of A</b> $ \mathbf{A}  =$	$\sqrt{A_x^2 + A_y^2 + A_z^2}$	$\sqrt{A_r^2 + A_\phi^2 + A_z^2}$	$\sqrt{A_R^2 + A_\theta^2 + A_\phi^2}$
<b>Position vector</b> $\overrightarrow{OP_1} =$	$\hat{\mathbf{x}}x_1 + \hat{\mathbf{y}}y_1 + \hat{\mathbf{z}}z_1,$ for $P(x_1, y_1, z_1)$	$\hat{\mathbf{r}}r_1 + \hat{\mathbf{z}}z_1,$ for $P(r_1, \phi_1, z_1)$	$\hat{\mathbf{R}}R_1,$ for $P(R_1, \theta_1, \phi_1)$
<b>Base vectors properties</b>	$\hat{\mathbf{x}} \cdot \hat{\mathbf{x}} = \hat{\mathbf{y}} \cdot \hat{\mathbf{y}} = \hat{\mathbf{z}} \cdot \hat{\mathbf{z}} = 1$ $\hat{\mathbf{x}} \cdot \hat{\mathbf{y}} = \hat{\mathbf{y}} \cdot \hat{\mathbf{z}} = \hat{\mathbf{z}} \cdot \hat{\mathbf{x}} = 0$ $\hat{\mathbf{x}} \times \hat{\mathbf{y}} = \hat{\mathbf{z}}$ $\hat{\mathbf{y}} \times \hat{\mathbf{z}} = \hat{\mathbf{x}}$ $\hat{\mathbf{z}} \times \hat{\mathbf{x}} = \hat{\mathbf{y}}$	$\hat{\mathbf{r}} \cdot \hat{\mathbf{r}} = \hat{\boldsymbol{\phi}} \cdot \hat{\boldsymbol{\phi}} = \hat{\mathbf{z}} \cdot \hat{\mathbf{z}} = 1$ $\hat{\mathbf{r}} \cdot \hat{\boldsymbol{\phi}} = \hat{\boldsymbol{\phi}} \cdot \hat{\mathbf{z}} = \hat{\mathbf{z}} \cdot \hat{\mathbf{r}} = 0$ $\hat{\mathbf{r}} \times \hat{\boldsymbol{\phi}} = \hat{\mathbf{z}}$ $\hat{\boldsymbol{\phi}} \times \hat{\mathbf{z}} = \hat{\mathbf{r}}$ $\hat{\mathbf{z}} \times \hat{\mathbf{r}} = \hat{\boldsymbol{\phi}}$	$\hat{\mathbf{R}} \cdot \hat{\mathbf{R}} = \hat{\boldsymbol{\theta}} \cdot \hat{\boldsymbol{\theta}} = \hat{\boldsymbol{\phi}} \cdot \hat{\boldsymbol{\phi}} = 1$ $\hat{\mathbf{R}} \cdot \hat{\boldsymbol{\theta}} = \hat{\boldsymbol{\theta}} \cdot \hat{\boldsymbol{\phi}} = \hat{\boldsymbol{\phi}} \cdot \hat{\mathbf{R}} = 0$ $\hat{\mathbf{R}} \times \hat{\boldsymbol{\theta}} = \hat{\boldsymbol{\phi}}$ $\hat{\boldsymbol{\theta}} \times \hat{\boldsymbol{\phi}} = \hat{\mathbf{R}}$ $\hat{\boldsymbol{\phi}} \times \hat{\mathbf{R}} = \hat{\boldsymbol{\theta}}$
<b>Dot product</b> $\mathbf{A} \cdot \mathbf{B} =$	$A_x B_x + A_y B_y + A_z B_z$	$A_r B_r + A_\phi B_\phi + A_z B_z$	$A_R B_R + A_\theta B_\theta + A_\phi B_\phi$
<b>Cross product</b> $\mathbf{A} \times \mathbf{B} =$	$\begin{vmatrix} \hat{\mathbf{x}} & \hat{\mathbf{y}} & \hat{\mathbf{z}} \\ A_x & A_y & A_z \\ B_x & B_y & B_z \end{vmatrix}$	$\begin{vmatrix} \hat{\mathbf{r}} & \hat{\boldsymbol{\phi}} & \hat{\mathbf{z}} \\ A_r & A_\phi & A_z \\ B_r & B_\phi & B_z \end{vmatrix}$	$\begin{vmatrix} \hat{\mathbf{R}} & \hat{\boldsymbol{\theta}} & \hat{\boldsymbol{\phi}} \\ A_R & A_\theta & A_\phi \\ B_R & B_\theta & B_\phi \end{vmatrix}$
<b>Differential length</b> $d\mathbf{l} =$	$\hat{\mathbf{x}} dx + \hat{\mathbf{y}} dy + \hat{\mathbf{z}} dz$	$\hat{\mathbf{r}} dr + \hat{\boldsymbol{\phi}} r d\phi + \hat{\mathbf{z}} dz$	$\hat{\mathbf{R}} dR + \hat{\boldsymbol{\theta}} R d\theta + \hat{\boldsymbol{\phi}} R \sin \theta d\phi$
<b>Differential surface areas</b>	$ds_x = \hat{\mathbf{x}} dy dz$ $ds_y = \hat{\mathbf{y}} dx dz$ $ds_z = \hat{\mathbf{z}} dx dy$	$ds_r = \hat{\mathbf{r}} r d\phi dz$ $ds_\phi = \hat{\boldsymbol{\phi}} dr dz$ $ds_z = \hat{\mathbf{z}} r dr d\phi$	$ds_R = \hat{\mathbf{R}} R^2 \sin \theta d\theta d\phi$ $ds_\theta = \hat{\boldsymbol{\theta}} R \sin \theta dR d\phi$ $ds_\phi = \hat{\boldsymbol{\phi}} R dR d\theta$
<b>Differential volume</b> $dV =$	$dx dy dz$	$r dr d\phi dz$	$R^2 \sin \theta dR d\theta d\phi$

**Table 3-2:** Coordinate transformation relations.

Transformation	Coordinate Variables	Unit Vectors	Vector Components
<b>Cartesian to cylindrical</b>	$r = \sqrt{x^2 + y^2}$ $\phi = \tan^{-1}(y/x)$ $z = z$	$\hat{\mathbf{r}} = \hat{\mathbf{x}} \cos \phi + \hat{\mathbf{y}} \sin \phi$ $\hat{\boldsymbol{\phi}} = -\hat{\mathbf{x}} \sin \phi + \hat{\mathbf{y}} \cos \phi$ $\hat{\mathbf{z}} = \hat{\mathbf{z}}$	$A_r = A_x \cos \phi + A_y \sin \phi$ $A_\phi = -A_x \sin \phi + A_y \cos \phi$ $A_z = A_z$
<b>Cylindrical to Cartesian</b>	$x = r \cos \phi$ $y = r \sin \phi$ $z = z$	$\hat{\mathbf{x}} = \hat{\mathbf{r}} \cos \phi - \hat{\boldsymbol{\phi}} \sin \phi$ $\hat{\mathbf{y}} = \hat{\mathbf{r}} \sin \phi + \hat{\boldsymbol{\phi}} \cos \phi$ $\hat{\mathbf{z}} = \hat{\mathbf{z}}$	$A_x = A_r \cos \phi - A_\phi \sin \phi$ $A_y = A_r \sin \phi + A_\phi \cos \phi$ $A_z = A_z$
<b>Cartesian to spherical</b>	$R = \sqrt{x^2 + y^2 + z^2}$ $\theta = \tan^{-1}[\sqrt{x^2 + y^2}/z]$ $\phi = \tan^{-1}(y/x)$	$\hat{\mathbf{R}} = \hat{\mathbf{x}} \sin \theta \cos \phi + \hat{\mathbf{y}} \sin \theta \sin \phi + \hat{\mathbf{z}} \cos \theta$ $\hat{\boldsymbol{\theta}} = \hat{\mathbf{x}} \cos \theta \cos \phi + \hat{\mathbf{y}} \cos \theta \sin \phi - \hat{\mathbf{z}} \sin \theta$ $\hat{\boldsymbol{\phi}} = -\hat{\mathbf{x}} \sin \phi + \hat{\mathbf{y}} \cos \phi$	$A_R = A_x \sin \theta \cos \phi + A_y \sin \theta \sin \phi + A_z \cos \theta$ $A_\theta = A_x \cos \theta \cos \phi + A_y \cos \theta \sin \phi - A_z \sin \theta$ $A_\phi = -A_x \sin \phi + A_y \cos \phi$
<b>Spherical to Cartesian</b>	$x = R \sin \theta \cos \phi$ $y = R \sin \theta \sin \phi$ $z = R \cos \theta$	$\hat{\mathbf{x}} = \hat{\mathbf{R}} \sin \theta \cos \phi + \hat{\boldsymbol{\theta}} \cos \theta \cos \phi - \hat{\boldsymbol{\phi}} \sin \phi$ $\hat{\mathbf{y}} = \hat{\mathbf{R}} \sin \theta \sin \phi + \hat{\boldsymbol{\theta}} \cos \theta \sin \phi + \hat{\boldsymbol{\phi}} \cos \phi$ $\hat{\mathbf{z}} = \hat{\mathbf{R}} \cos \theta - \hat{\boldsymbol{\theta}} \sin \theta$	$A_x = A_R \sin \theta \cos \phi + A_\theta \cos \theta \cos \phi - A_\phi \sin \phi$ $A_y = A_R \sin \theta \sin \phi + A_\theta \cos \theta \sin \phi + A_\phi \cos \phi$ $A_z = A_R \cos \theta - A_\theta \sin \theta$
<b>Cylindrical to spherical</b>	$R = \sqrt{r^2 + z^2}$ $\theta = \tan^{-1}(r/z)$ $\phi = \phi$	$\hat{\mathbf{R}} = \hat{\mathbf{r}} \sin \theta + \hat{\mathbf{z}} \cos \theta$ $\hat{\boldsymbol{\theta}} = \hat{\mathbf{r}} \cos \theta - \hat{\mathbf{z}} \sin \theta$ $\hat{\boldsymbol{\phi}} = \hat{\boldsymbol{\phi}}$	$A_R = A_r \sin \theta + A_z \cos \theta$ $A_\theta = A_r \cos \theta - A_z \sin \theta$ $A_\phi = A_\phi$
<b>Spherical to cylindrical</b>	$r = R \sin \theta$ $\phi = \phi$ $z = R \cos \theta$	$\hat{\mathbf{r}} = \hat{\mathbf{R}} \sin \theta + \hat{\boldsymbol{\theta}} \cos \theta$ $\hat{\boldsymbol{\phi}} = \hat{\boldsymbol{\phi}}$ $\hat{\mathbf{z}} = \hat{\mathbf{R}} \cos \theta - \hat{\boldsymbol{\theta}} \sin \theta$	$A_r = A_R \sin \theta + A_\theta \cos \theta$ $A_\phi = A_\phi$ $A_z = A_R \cos \theta - A_\theta \sin \theta$



## Chapter 4 Tables

**Table 4-1** Conductivity of some common materials at 20°C.

**Table 4-2** Relative permittivity (dielectric constant) and dielectric strength of common materials.

**Table 4-3** Boundary conditions for the electric fields.

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**Table 4-1:** Conductivity of some common materials at 20°C.

Material	Conductivity, $\sigma$ (S/m)
<i>Conductors</i>	
Silver	$6.2 \times 10^7$
Copper	$5.8 \times 10^7$
Gold	$4.1 \times 10^7$
Aluminum	$3.5 \times 10^7$
Iron	$10^7$
Mercury	$10^6$
Carbon	$3 \times 10^4$
<i>Semiconductors</i>	
Pure germanium	2.2
Pure silicon	$4.4 \times 10^{-4}$
<i>Insulators</i>	
Glass	$10^{-12}$
Paraffin	$10^{-15}$
Mica	$10^{-15}$
Fused quartz	$10^{-17}$

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**Table 4-2:** Relative permittivity (dielectric constant) and dielectric strength of common materials.

Material	Relative Permittivity, $\epsilon_r$	Dielectric Strength, $E_{ds}$ (MV/m)
Air (at sea level)	1.0006	3
Petroleum oil	2.1	12
Polystyrene	2.6	20
Glass	4.5–10	25–40
Quartz	3.8–5	30
Bakelite	5	20
Mica	5.4–6	200

$$\epsilon = \epsilon_r \epsilon_0 \text{ and } \epsilon_0 = 8.854 \times 10^{-12} \text{ F/m.}$$

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**Table 4-3:** Boundary conditions for the electric fields.

Field Component	Any Two Media	Medium 1 Dielectric $\epsilon_1$	Medium 2 Conductor
<b>Tangential E</b>	$\mathbf{E}_{1t} = \mathbf{E}_{2t}$	$\mathbf{E}_{1t} = \mathbf{E}_{2t} = 0$	
<b>Tangential D</b>	$\mathbf{D}_{1t}/\epsilon_1 = \mathbf{D}_{2t}/\epsilon_2$	$\mathbf{D}_{1t} = \mathbf{D}_{2t} = 0$	
<b>Normal E</b>	$\epsilon_1 E_{1n} - \epsilon_2 E_{2n} = \rho_s$	$E_{1n} = \rho_s/\epsilon_1$	$E_{2n} = 0$
<b>Normal D</b>	$D_{1n} - D_{2n} = \rho_s$	$D_{1n} = \rho_s$	$D_{2n} = 0$

Notes: (1)  $\rho_s$  is the surface charge density at the boundary; (2) normal components of  $\mathbf{E}_1$ ,  $\mathbf{D}_1$ ,  $\mathbf{E}_2$ , and  $\mathbf{D}_2$  are along  $\hat{\mathbf{n}}_2$ , the outward normal unit vector of medium 2.

## Chapter 5 Tables

**Table 5-1** Attributes of electrostatics and magnetostatics.

**Table 5-2** Properties of magnetic materials.

**Table 5-1:** Attributes of electrostatics and magnetostatics.

Attribute	Electrostatics	Magnetostatics
<b>Sources</b>	Stationary charges $\rho_v$	Steady currents $\mathbf{J}$
<b>Fields and Fluxes</b>	$\mathbf{E}$ and $\mathbf{D}$	$\mathbf{H}$ and $\mathbf{B}$
<b>Constitutive parameter(s)</b>	$\epsilon$ and $\sigma$	$\mu$
<b>Governing equations</b>		
• <b>Differential form</b>	$\nabla \cdot \mathbf{D} = \rho_v$ $\nabla \times \mathbf{E} = 0$	$\nabla \cdot \mathbf{B} = 0$ $\nabla \times \mathbf{H} = \mathbf{J}$
• <b>Integral form</b>	$\oint_S \mathbf{D} \cdot d\mathbf{s} = Q$ $\oint_C \mathbf{E} \cdot d\mathbf{l} = 0$	$\oint_S \mathbf{B} \cdot d\mathbf{s} = 0$ $\oint_C \mathbf{H} \cdot d\mathbf{l} = I$
<b>Potential</b>	Scalar $V$ , with $\mathbf{E} = -\nabla V$	Vector $\mathbf{A}$ , with $\mathbf{B} = \nabla \times \mathbf{A}$
<b>Energy density</b>	$w_e = \frac{1}{2} \epsilon E^2$	$w_m = \frac{1}{2} \mu H^2$
<b>Force on charge <math>q</math></b>	$\mathbf{F}_e = q\mathbf{E}$	$\mathbf{F}_m = q\mathbf{u} \times \mathbf{B}$
<b>Circuit element(s)</b>	$C$ and $R$	$L$

**Table 5-2:** Properties of magnetic materials.

	<b>Diamagnetism</b>	<b>Paramagnetism</b>	<b>Ferromagnetism</b>
<b>Permanent magnetic dipole moment</b>	No	Yes, but weak	Yes, and strong
<b>Primary magnetization mechanism</b>	Electron orbital magnetic moment	Electron spin magnetic moment	Magnetized domains
<b>Direction of induced magnetic field (relative to external field)</b>	Opposite	Same	Hysteresis [see Fig. ??]
<b>Common substances</b>	Bismuth, copper, diamond, gold, lead, mercury, silver, silicon	Aluminum, calcium, chromium, magnesium, niobium, platinum, tungsten	Iron, nickel, cobalt
<b>Typical value of <math>\chi_m</math></b> <b>Typical value of <math>\mu_r</math></b>	$\approx -10^{-5}$ $\approx 1$	$\approx 10^{-5}$ $\approx 1$	$ \chi_m  \gg 1$ and hysteretic $ \mu_r  \gg 1$ and hysteretic

## Chapter 6 Tables

**Table 6-1** Maxwell's equations.

**Table 6-2** Boundary conditions for the electric and magnetic fields.



**Table 6-1:** Maxwell's equations.

Reference	Differential Form	Integral Form
<b>Gauss's law</b>	$\nabla \cdot \mathbf{D} = \rho_v$	$\oint_S \mathbf{D} \cdot d\mathbf{s} = Q$ (6.1)
<b>Faraday's law</b>	$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t}$	$\oint_C \mathbf{E} \cdot d\mathbf{l} = -\int_S \frac{\partial \mathbf{B}}{\partial t} \cdot d\mathbf{s}$ (6.2)*
<b>No magnetic charges (Gauss's law for magnetism)</b>	$\nabla \cdot \mathbf{B} = 0$	$\oint_S \mathbf{B} \cdot d\mathbf{s} = 0$ (6.3)
<b>Ampère's law</b>	$\nabla \times \mathbf{H} = \mathbf{J} + \frac{\partial \mathbf{D}}{\partial t}$	$\oint_C \mathbf{H} \cdot d\mathbf{l} = \int_S \left( \mathbf{J} + \frac{\partial \mathbf{D}}{\partial t} \right) \cdot d\mathbf{s}$ (6.4)
*For a stationary surface $S$ .		

**Table 6-2:** Boundary conditions for the electric and magnetic fields.

Field Components	General Form	Medium 1 Dielectric	Medium 2 Dielectric	Medium 1 Dielectric	Medium 2 Conductor
<b>Tangential E</b>	$\hat{\mathbf{n}}_2 \times (\mathbf{E}_1 - \mathbf{E}_2) = 0$		$E_{1t} = E_{2t}$		$E_{1t} = E_{2t} = 0$
<b>Normal D</b>	$\hat{\mathbf{n}}_2 \cdot (\mathbf{D}_1 - \mathbf{D}_2) = \rho_s$		$D_{1n} - D_{2n} = \rho_s$		$D_{2n} = 0$
<b>Tangential H</b>	$\hat{\mathbf{n}}_2 \times (\mathbf{H}_1 - \mathbf{H}_2) = \mathbf{J}_s$		$H_{1t} = H_{2t}$		$H_{2t} = 0$
<b>Normal B</b>	$\hat{\mathbf{n}}_2 \cdot (\mathbf{B}_1 - \mathbf{B}_2) = 0$		$B_{1n} = B_{2n}$		$B_{1n} = B_{2n} = 0$
Notes: (1) $\rho_s$ is the surface charge density at the boundary; (2) $\mathbf{J}_s$ is the surface current density at the boundary; (3) normal components of all fields are along $\hat{\mathbf{n}}_2$ , the outward unit vector of medium 2; (4) $E_{1t} = E_{2t}$ implies that the tangential components are equal in magnitude and parallel in direction; (5) direction of $\mathbf{J}_s$ is orthogonal to $(\mathbf{H}_1 - \mathbf{H}_2)$ .					

## Chapter 7 Tables

**Table 7-1** Expressions for  $\alpha$ ,  $\beta$ ,  $\eta_c$ ,  $u_p$ , and  $\lambda$  for various types of media.

**Table 7-2** Power ratios in natural numbers and in decibels.

**Table 7-1:** Expressions for  $\alpha$ ,  $\beta$ ,  $\eta_c$ ,  $u_p$ , and  $\lambda$  for various types of media.

	Any Medium	Lossless Medium ( $\sigma = 0$ )	Low-loss Medium ( $\epsilon''/\epsilon' \ll 1$ )	Good Conductor ( $\epsilon''/\epsilon' \gg 1$ )	Units
$\alpha =$	$\omega \left[ \frac{\mu\epsilon'}{2} \left[ \sqrt{1 + \left(\frac{\epsilon''}{\epsilon'}\right)^2} - 1 \right] \right]^{1/2}$	0	$\frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}}$	$\sqrt{\pi f \mu \sigma}$	(Np/m)
$\beta =$	$\omega \left[ \frac{\mu\epsilon'}{2} \left[ \sqrt{1 + \left(\frac{\epsilon''}{\epsilon'}\right)^2} + 1 \right] \right]^{1/2}$	$\omega\sqrt{\mu\epsilon}$	$\omega\sqrt{\mu\epsilon}$	$\sqrt{\pi f \mu \sigma}$	(rad/m)
$\eta_c =$	$\sqrt{\frac{\mu}{\epsilon'}} \left( 1 - j \frac{\epsilon''}{\epsilon'} \right)^{-1/2}$	$\sqrt{\frac{\mu}{\epsilon}}$	$\sqrt{\frac{\mu}{\epsilon}}$	$(1 + j) \frac{\alpha}{\sigma}$	( $\Omega$ )
$u_p =$	$\omega/\beta$	$1/\sqrt{\mu\epsilon}$	$1/\sqrt{\mu\epsilon}$	$\sqrt{4\pi f/\mu\sigma}$	(m/s)
$\lambda =$	$2\pi/\beta = u_p/f$	$u_p/f$	$u_p/f$	$u_p/f$	(m)
Notes: $\epsilon' = \epsilon$ ; $\epsilon'' = \sigma/\omega$ ; in free space, $\epsilon = \epsilon_0$ , $\mu = \mu_0$ ; in practice, a material is considered a low-loss medium if $\epsilon''/\epsilon' = \sigma/\omega\epsilon < 0.01$ and a good conducting medium if $\epsilon''/\epsilon' > 100$ .					

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**Table 7-2:** Power ratios in natural numbers and in decibels.

$G$	$G$ [dB]
$10^x$	$10x$ dB
4	6 dB
2	3 dB
1	0 dB
0.5	-3 dB
0.25	-6 dB
0.1	-10 dB
$10^{-3}$	-30 dB

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## Chapter 8 Tables

**Table 8-1** Analogy between plane-wave equations for normal incidence and transmission-line equations, both under lossless conditions.

**Table 8-1:** Analogy between plane-wave equations for normal incidence and transmission-line equations, both under lossless conditions.

Plane Wave [Fig. ??(a)]	Transmission Line [Fig. ??(b)]
$\tilde{\mathbf{E}}_1(z) = \hat{\mathbf{x}}E_0^i(e^{-jk_1z} + \Gamma e^{jk_1z})$ (8.5a)	$\tilde{V}_1(z) = V_0^+(e^{-j\beta_1z} + \Gamma e^{j\beta_1z})$ (8.5b)
$\tilde{\mathbf{H}}_1(z) = \hat{\mathbf{y}} \frac{E_0^i}{\eta_1}(e^{-jk_1z} - \Gamma e^{jk_1z})$ (8.6a)	$\tilde{I}_1(z) = \frac{V_0^+}{Z_{01}}(e^{-j\beta_1z} - \Gamma e^{j\beta_1z})$ (8.6b)
$\tilde{\mathbf{E}}_2(z) = \hat{\mathbf{x}}\tau E_0^i e^{-jk_2z}$ (8.7a)	$\tilde{V}_2(z) = \tau V_0^+ e^{-j\beta_2z}$ (8.7b)
$\tilde{\mathbf{H}}_2(z) = \hat{\mathbf{y}}\tau \frac{E_0^i}{\eta_2} e^{-jk_2z}$ (8.8a)	$\tilde{I}_2(z) = \tau \frac{V_0^+}{Z_{02}} e^{-j\beta_2z}$ (8.8b)
$\Gamma = (\eta_2 - \eta_1)/(\eta_2 + \eta_1)$	$\Gamma = (Z_{02} - Z_{01})/(Z_{02} + Z_{01})$
$\tau = 1 + \Gamma$	$\tau = 1 + \Gamma$
$k_1 = \omega\sqrt{\mu_1\epsilon_1}, \quad k_2 = \omega\sqrt{\mu_2\epsilon_2}$	$\beta_1 = \omega\sqrt{\mu_1\epsilon_1}, \quad \beta_2 = \omega\sqrt{\mu_2\epsilon_2}$
$\eta_1 = \sqrt{\mu_1/\epsilon_1}, \quad \eta_2 = \sqrt{\mu_2/\epsilon_2}$	$Z_{01}$ and $Z_{02}$ depend on transmission-line parameters

## Chapter 9 Tables

There are no Tables in Chapter 9.



## Chapter 10 Tables

**Table 10-1** Communications satellite frequency allocations.

**Table 10-1:** Communications satellite frequency allocations.

<b>Use</b>	<b>Downlink Frequency (MHz)</b>	<b>Uplink Frequency (MHz)</b>
<b>Fixed Service</b>		
Commercial (C-band)	3,700–4,200	5,925–6,425
Military (X-band)	7,250–7,750	7,900–8,400
Commercial (K-band)		
Domestic (USA)	11,700–12,200	14,000–14,500
International	10,950–11,200	27,500–31,000
<b>Mobile Service</b>		
Maritime	1,535–1,542.5	1,635–1,644
Aeronautical	1,543.5–1,558.8	1,645–1,660
<b>Broadcast Service</b>		
	2,500–2,535	2,655–2,690
	11,700–12,750	
<b>Telemetry, Tracking, and Command</b>		
	137–138, 401–402, 1,525–1,540	